

**OVER CURRENT PROTECTION IC  
WITH SHUNT REGULATOR****GENERAL DESCRIPTION**

**FP133** is a main rail current detection and over current protection IC. It is including a current shunt comparator and shutdown comparator with a precision shunt regulator like **FP431**.

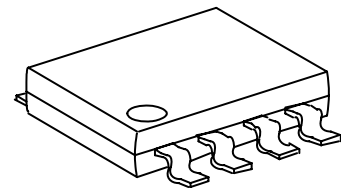
The current shunt comparator has the any gain adjustable function for rail current detection using three external resistors, and inter-connecting to a shutdown comparator for driving a protection circuit like a photo-coupler to shutdown the primary side PWM IC when over current is occurred.

And the voltage shunt regulator has a **2.5V** reference for switching power supply secondary output voltage feedback.

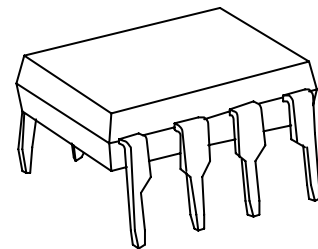
**FP133** is used with a few external parts for O.C.P. and output voltage feedback function and good active point during wide temperature range in secondary main rail power supply of SPS or isolated fly-back DC-DC converter application field.

**FEATURES**

- Wide operating voltage range: +2.7V~ +30V
- Wide operating temperature range: -20°C~ +105°C
- Independent shunt and supply voltage
- Low input offset voltage
- Sense gain adjustable
- Built-in 1.25V comparator for O.C.P.
- Shunt regulator voltage: 2.5V (0.5%)
- FPO Output sink current capability up 40mA
- Package: PDIP8/SOP8



SOP8

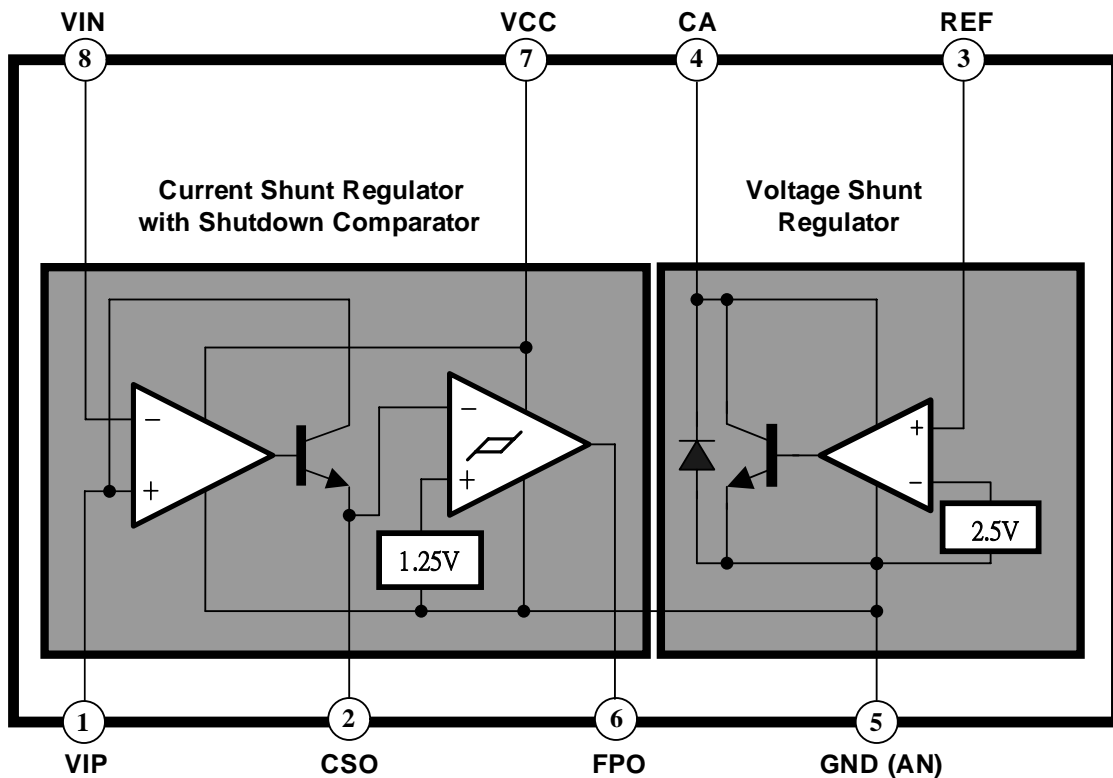


PDIP8

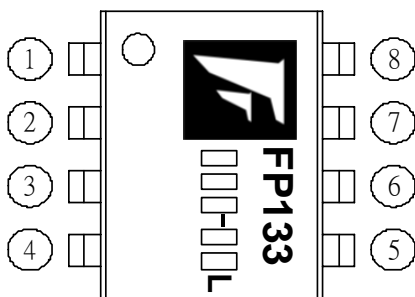
**TYPICAL APPLICATION**

- SPS
- AC Adaptor
- Isolated Fly-back DC-DC Converter

## FUNCTIONAL BLOCK DIAGRAM



### MARK VIEW



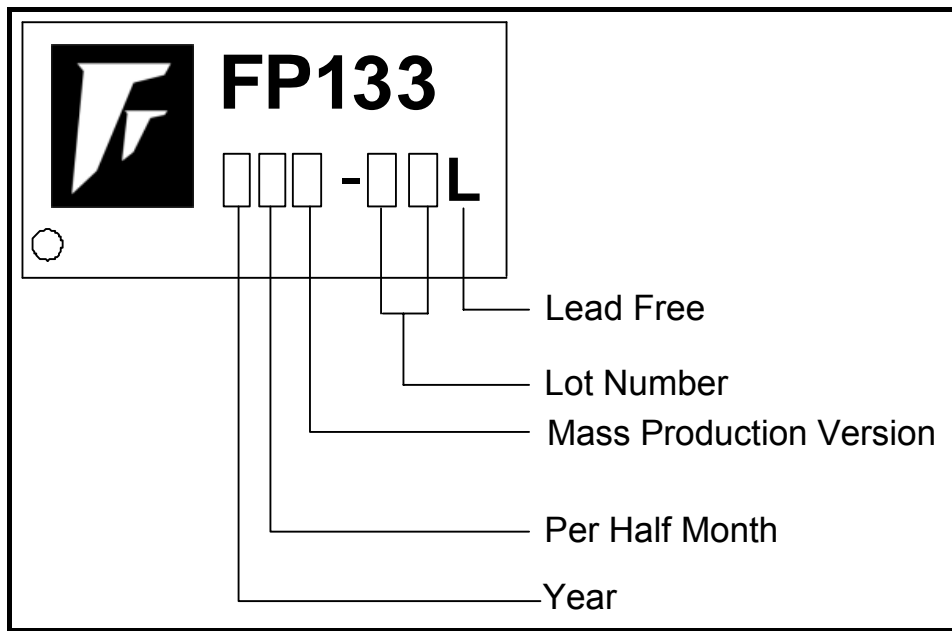
### PIN DESCRIPTION

Name	No.	Status	Description
VIP	1	I	Positive Input of Current Shunt OPA
CSO	2	O/I	Output of Current Shunt OPA to Inverting of Shutdown Comparator
REF	3	I	2.5V Reference
CA	4	I	Voltage Shunt Cathode Input
GND	5	P	IC Ground
FPO	6	O	Shutdown Comparator Output (O.C.)
VCC	7	P	IC Power Supply
VIN	8	I	Negative Input of Current Shunt OPA

## ORDER INFORMATION

Part Number	Operating Temperature	Package	Description
FP133P-LF	-20°C ~ +105°C	PDIP8	Tube
FP133D-LF	-20°C ~ +105°C	SOP8	Tube
FP133DR-LF	-20°C ~ +105°C	SOP8	Tape & Reel

## IC DATE CODE DISTINGUISH



### FOR EXAMPLE:

January      A (Front Half Month), B (Last Half Month)

February     C, D

March        E, F            -----And so on.

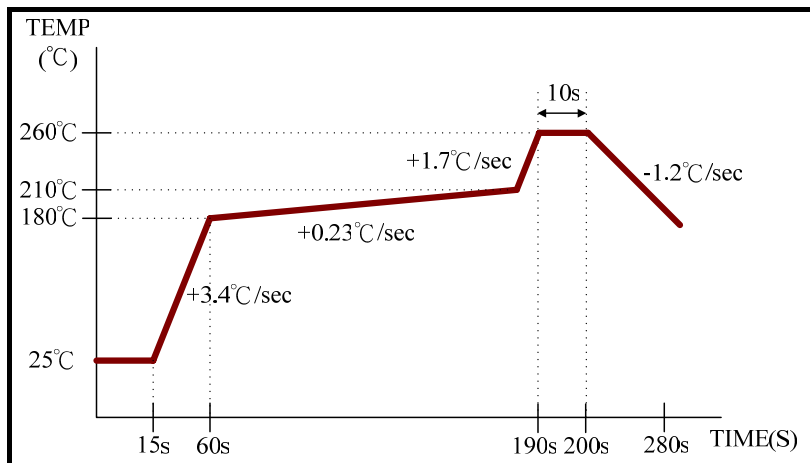
Lot Number is the last two numbers

### For Example:

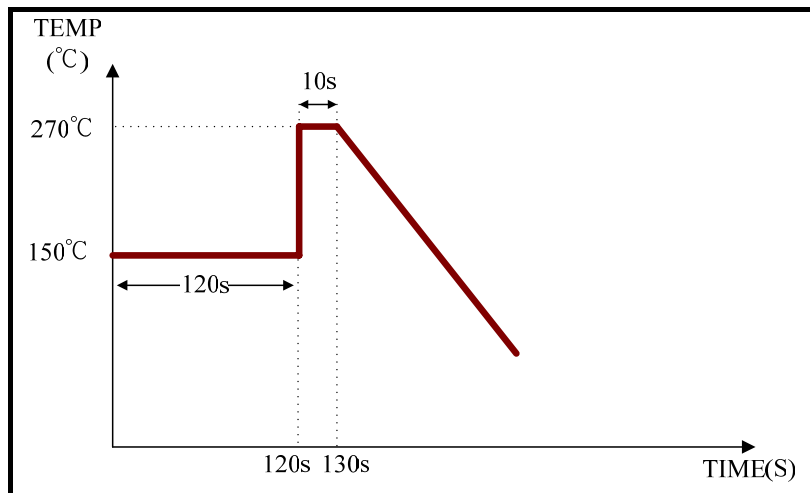
A3311C62  
 ↳ Lot Number

## ABSOLUTE MAXIMUM RATINGS

Power Supply Voltage (VCC) -----	+30V
Current Shunt Regulator Common Mode Inputs Voltage -----	-0.3V ~ 30V
Current Shunt Regulator differential Inputs Voltage (VIP – VIN) -----	-30V ~ 1.5V
CSO Voltage -----	-0.3V ~ V <sub>CC</sub>
FPO Sink Current -----	+40mA
FPO Off Voltage -----	+30V
Cathode Voltage -----	+30V
Cathode Continuous Current-----	-100mA~+150mA
Reference Input Current -----	-50 $\mu$ A~+1mA
Operating Junction Temperature (T <sub>J</sub> )-----	+150 $^{\circ}$ C
Operating Junction Range (T <sub>A</sub> )-----	-20 $^{\circ}$ C ~ +105 $^{\circ}$ C
Storage Temperature Range (T <sub>s</sub> ) -----	-55 $^{\circ}$ C ~ +150 $^{\circ}$ C
Power Dissipation (SOP8, T <sub>A</sub> =25 $^{\circ}$ C) -----	570mW
(P-DIP8, T <sub>A</sub> =25 $^{\circ}$ C)-----	900mW
SOP8 Lead Temperature (IR Re-flow soldering , 10 sec) -----	+260 $^{\circ}$ C
PDIP8 Lead Temperature (Wave soldering , 10 sec) -----	+270 $^{\circ}$ C



IR Re-flow Temperature vs. Second Curve



Wave Soldering Temperature vs. Second Cure

## DC ELECTRICAL CHARACTERISTICS

( $T_A = -20^{\circ}\text{C} \sim +105^{\circ}\text{C}$ ,  $V_{CC} = V_{IP} = 12\text{V}$ ,  $R_{G1} = R_{G2} = 5\text{K}$ ,  $R_L = 125\text{K}\Omega$ , unless otherwise noted)

### Current Shunt Comparator section

PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
Full Scale Sense Voltage	$V_{SENSE}$	$V_{SENSE} = V_{IP} - V_{IN}$		100	500	mV
Common-Mode Input Voltage	$V_{CM}$		2.7		30	V
Common-Mode Rejection	CMRR	$V_{IP} = 2.7\text{V}$ to $30\text{V}$ , $V_{SENSE} = 50\text{mV}$	100	120		dB
Input Offset Voltage vs temp	$V_{OFFSET(ta)}$	$-20^{\circ}\text{C}$ to $+105^{\circ}\text{C}$		4	10	$\mu\text{V}/^{\circ}\text{C}$
Input Offset Voltage vs $V_{CC}$	$V_{OFFSET(vcc)}$	$V_{IN} = 2.7\text{V}$ to $30\text{V}$ , $V_{SENSE} = 50\text{mV}$		2.5	10	$\mu\text{V}/\text{V}$
Input Bias Current	$I_{BIAS}$	$V_{IP} = V_{IN} = 12\text{V}$		2		$\mu\text{A}$
Non-linearity Error	NLE	$V_{SENSE} = 10\text{mV}$ to $150\text{mV}$			$\pm 2$	%
Total Output Error	TOE	$V_{SENSE} = 100\text{mV}$			$\pm 2$	%
Output Impedance	$R_L$	-		1  5		$\text{G}\Omega    \text{pF}$
Voltage Swing to $V_{CC}$	$V_{SCC}$	-		$V_{CC} - 0.8$		V
Voltage Swing to $V_{CM}$	$V_{SCM}$	-		$V_{CM} - 0.8$		V
Bandwidth	BW	$R_{G1} = R_{G2} = 5\text{K}$ $R_L = 125\text{K}\Omega$		32		kHz
Settling Time	$t_s$	$R_{G1} = R_{G2} = 5\text{K}$ 5V Step, $R_L = 125\text{K}\Omega$		30		$\mu\text{s}$
Total Output-Current Noise	$I_{NOISE}$	BW=100KHz		3		nA
Operating Voltage Range	$V_{CC}$	-	2.7		30	V
Operating Temperature Range	$T_A$	-	-20		+105	$^{\circ}\text{C}$

### Shutdown Comparator section

PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
Common-Mode Voltage (CSO)	$V_{CM}$	-	-0.3	-	$V_{CC} - 1.5$	V
Voltage Gain	$A_V$	-	50	200	-	V/mV
Large Signal Response Time	$T_R$	-	-	300	-	nS
Response Time	$T_r$	-	-	1.3	-	$\mu\text{s}$
FPO Sink Current	$I_{SINK}$	$V_{CSO} > 1.3\text{V}$ $V_{FPO} \geq 1.0\text{V}$	20	-	-	mA
FPO Saturation Voltage	$V_{SAT}$	$V_{CSO} > 1.3\text{V}$ $I_{SINK} \leq 10\text{mA}$	-	-	1	V
FPO Leakage Current	$I_{L(FPO)}$	$V_{CSO} < 1.0\text{V}$ $V_{FPO} = 30\text{V}$	-	0.1	1	$\mu\text{A}$
CSO Threshold Voltage ( $V_{FPO}$ Switch High to Low)	$V_{TH}$	$T_A = 25^{\circ}\text{C}$	1.238	1.25	1.263	V
		$T_A = -20^{\circ}\text{C} \sim +105^{\circ}\text{C}$	1.225		1.275	
Hysteresis Voltage	$V_{HYS}$	$V_{FPO}$ Switch High to Low		100		mV

## DC ELECTRICAL CHARACTERISTICS (Cont.)

Test conditions:  $T_a = -20^{\circ}\text{C} \sim +105^{\circ}\text{C}$ ,  $V_{CC} = 12\text{V}$ , unless otherwise noted

### Voltage Shunt Regulator section

PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
Reference Voltage	$V_{REF}$	$V_{KA} = V_{REF}, I_{KA} = 10\text{mA}$	2.487	2.5	2.513	V
V Reference vs. Temperature	$V_{REF}$	$V_{KA} = V_{REF}, I_{KA} = 10\text{mA}$ $T_A = -20^{\circ}\text{C} \sim +105^{\circ}\text{C}$			30	mV
Line Regulation	$\Delta V_{REF} / \Delta V_{KA}$	$I_{KA} = 10\text{mA}$	$V_{KA} = 10\text{V} \sim V_{REF}$	-1.4	-2.0	mV/V
			$V_{KA} = 30\text{V} \sim 10\text{V}$	-1.0	-2.0	
Reference Current	$I_{REF}$	$R1 = 10\text{K}\Omega, R2 = \infty$ $I_{KA} = 10\text{mA}$		0.5	4.0	$\mu\text{A}$
I Reference vs. Temperature	$\Delta I_{REF}$	$R1 = 10\text{K}\Omega, R2 = \infty$ $I_{KA} = 10\text{mA}$ , $T_a = \text{Full range}$		0.4	1.2	$\mu\text{A}$
Minimum Cathode Current for Regulation	$I_{KA(MIN)}$	$V_{KA} = V_{REF}$		0.1	0.5	mA
Dynamic Impedance	$ Z_{KA} $	$V_{KA} = V_{REF}$ , $\Delta I_{KA} = 1\text{mA} \sim 100\text{mA}$ Frequency < 1KHz		0.2	0.5	$\Omega$

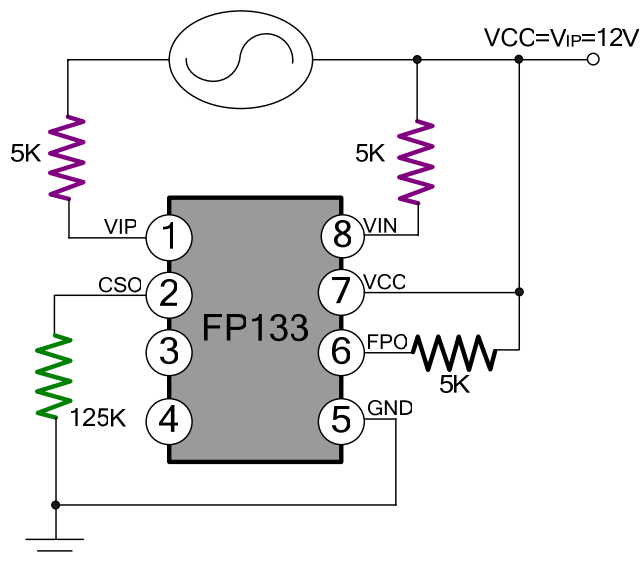
### Total device section

PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
Output Off IC Current	$I_{CC}$	$V_{CC} = 30\text{V}$	-	0.4	-	mA

## AC ELECTRICAL CHARACTERISTICS

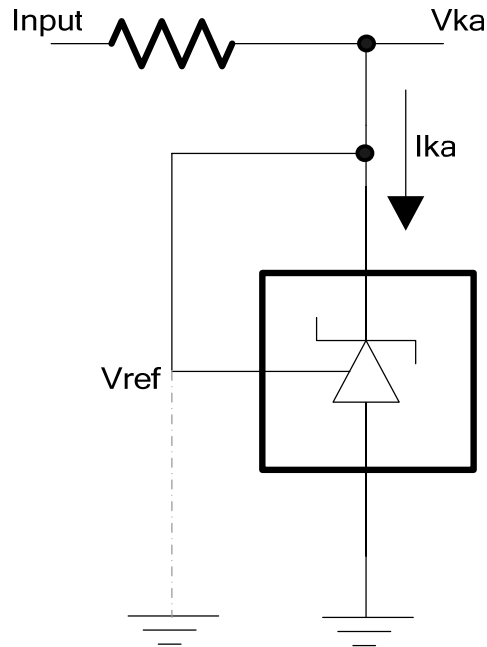
PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT
FPO Falling Delay Time	$t_{fd}$	See Test Circuit		6.4		$\mu\text{S}$
FPO Rising Delay Time	$t_{rd}$	See Test Circuit		4.4		$\mu\text{S}$

## PARAMETER MEASUREMENT INFORMATION

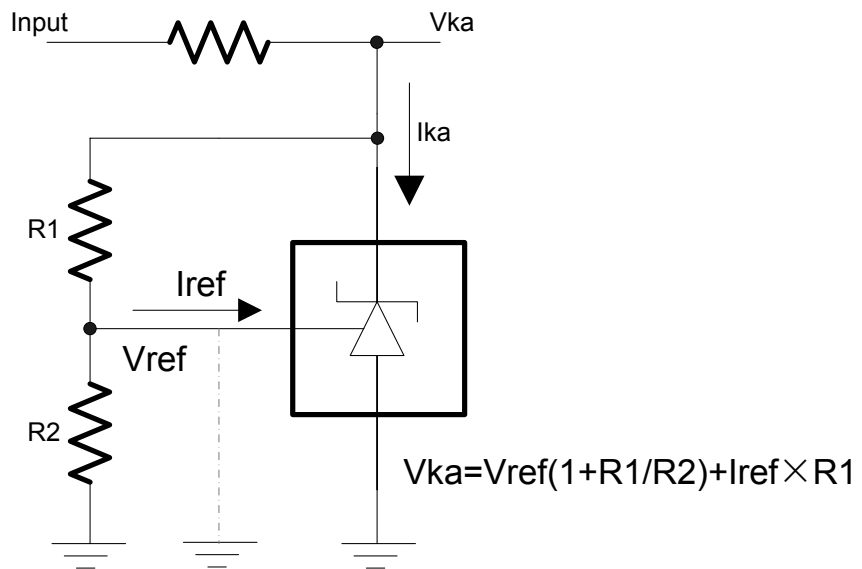


Test Circuit

PARAMETER MEASUREMENT INFORMATION(cont.)



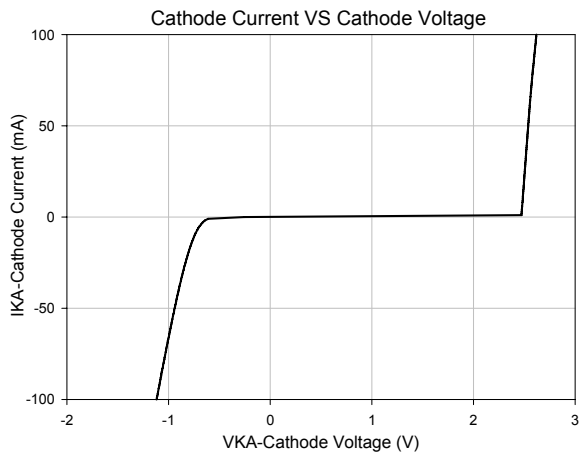
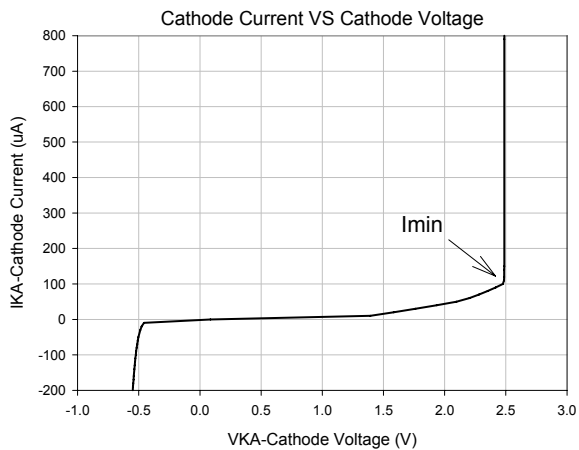
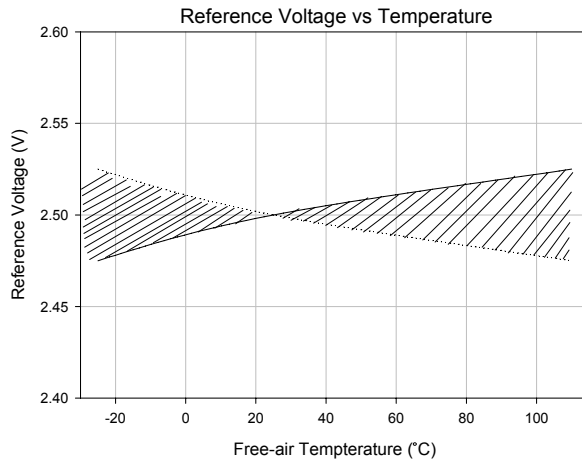
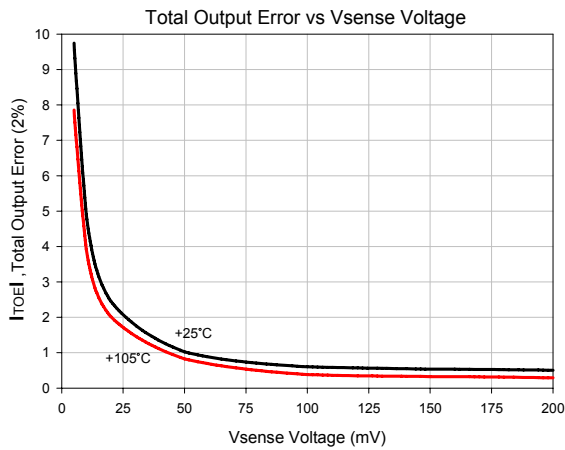
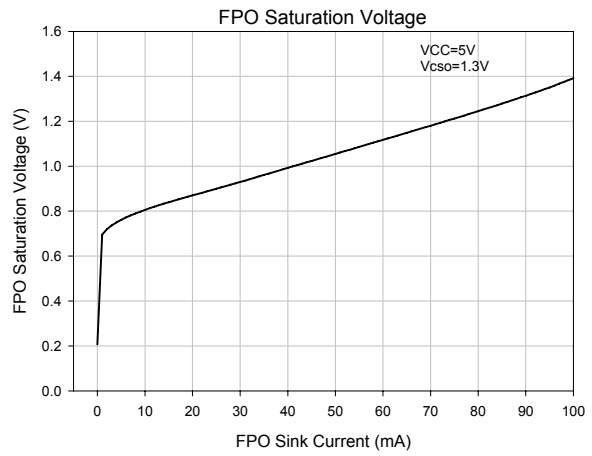
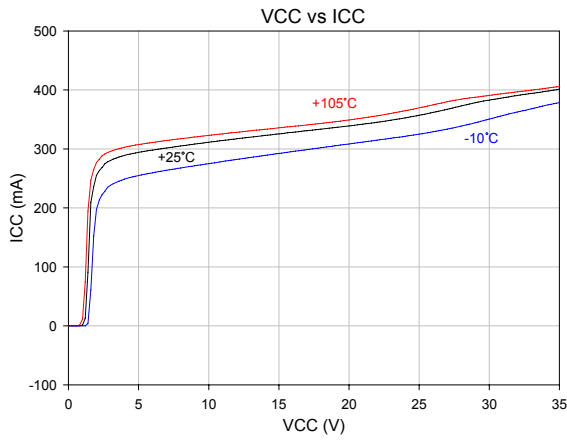
**Test Circuit for  $V_{ka}=V_{ref}$**



$$V_{ka} = V_{ref}(1 + R1/R2) + I_{ref} \times R1$$

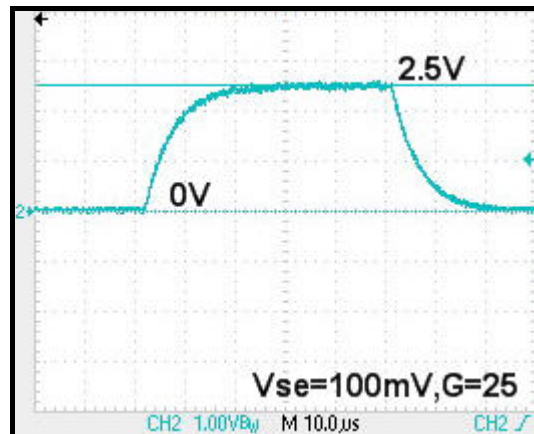
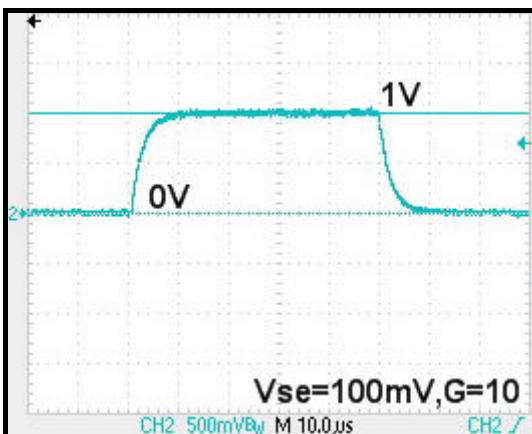
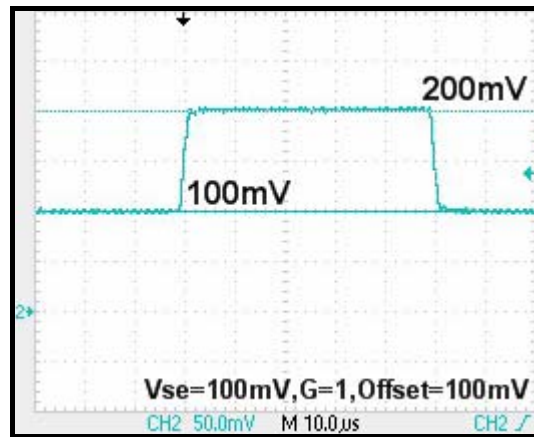
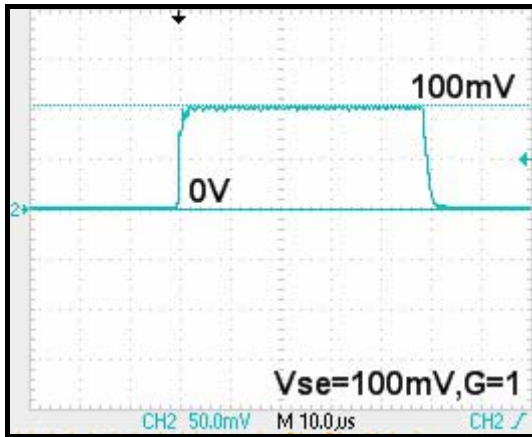
**Test Circuit for  $V_{ka} > V_{ref}$**

## TYPICAL CHARACTERISTICS (Ta=+25°C, VCC=12V, VIP=12V, unless otherwise noted)

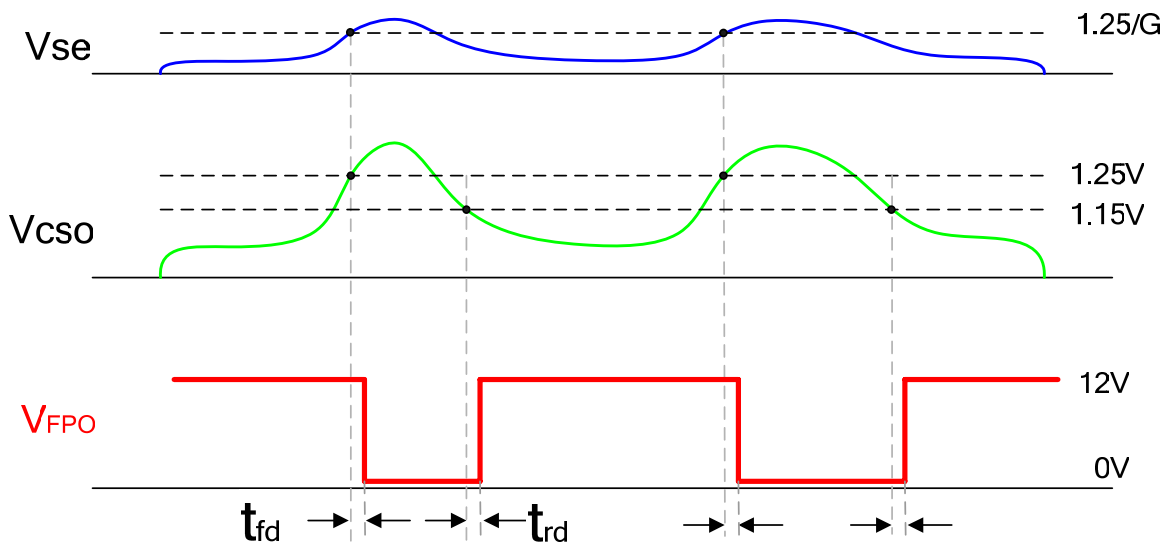


**TYPICAL CHARACTERISTICS** ( $T_a=+25^{\circ}\text{C}$ ,  $V_{CC}=12\text{V}$ ,  $V_{IP}=12\text{V}$ , unless otherwise noted)

$R_{G1}=R_{G2}=5\text{K}$



## Time Waveform Chart



## DETAILED DESCRIPTION

### Current Shunt Regulator

The figure shows the **FP133** current shunt block, when the load current ( $I_S$ ) flows from power supply and a dropout voltage ( $V_{IN}^+ - V_{IN}^-$ ) at the sense resistor ( $R_S$ ).

Assume internal NPN transistor collector current is same as emitter current ( $I_O$ ) and  $V_{IP}$  is very close  $V_{IN}$ , the **FP133** transfer function is:

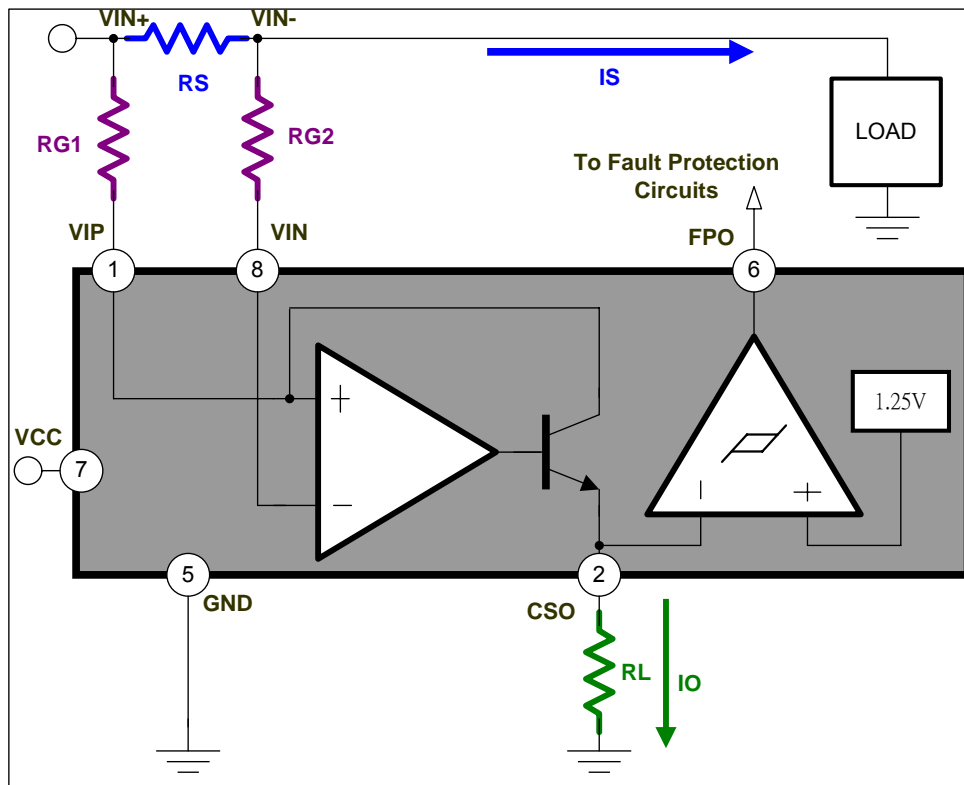
$$I_O = \frac{V_{IN}^+ - V_{IN}^-}{RG1} \quad \text{---- (1)}$$

In the figure, the ( $V_{IN}^+ - V_{IN}^-$ ), is equal to  $I_S \cdot R_S$  and the current shunt output voltage ( $CSO$ ) is equal to  $I_O \cdot R_L$ . The final transfer function for rail current measurement in this application is:

$$V_{CSO} = G * I_S * R_S \quad \text{---- (2)}$$

$$G = R_L / RG1 \quad \text{---- (3)}$$

In **FP133** internal circuits, the  $CSO$  output is connected with inverting input of shutdown comparator, when the voltage of  $CSO$  is higher than the internal reference voltage (**1.25V**), the  $FPO$  pin is switching from high to low state, this signal can be used for OLP protection control.



### NOTE

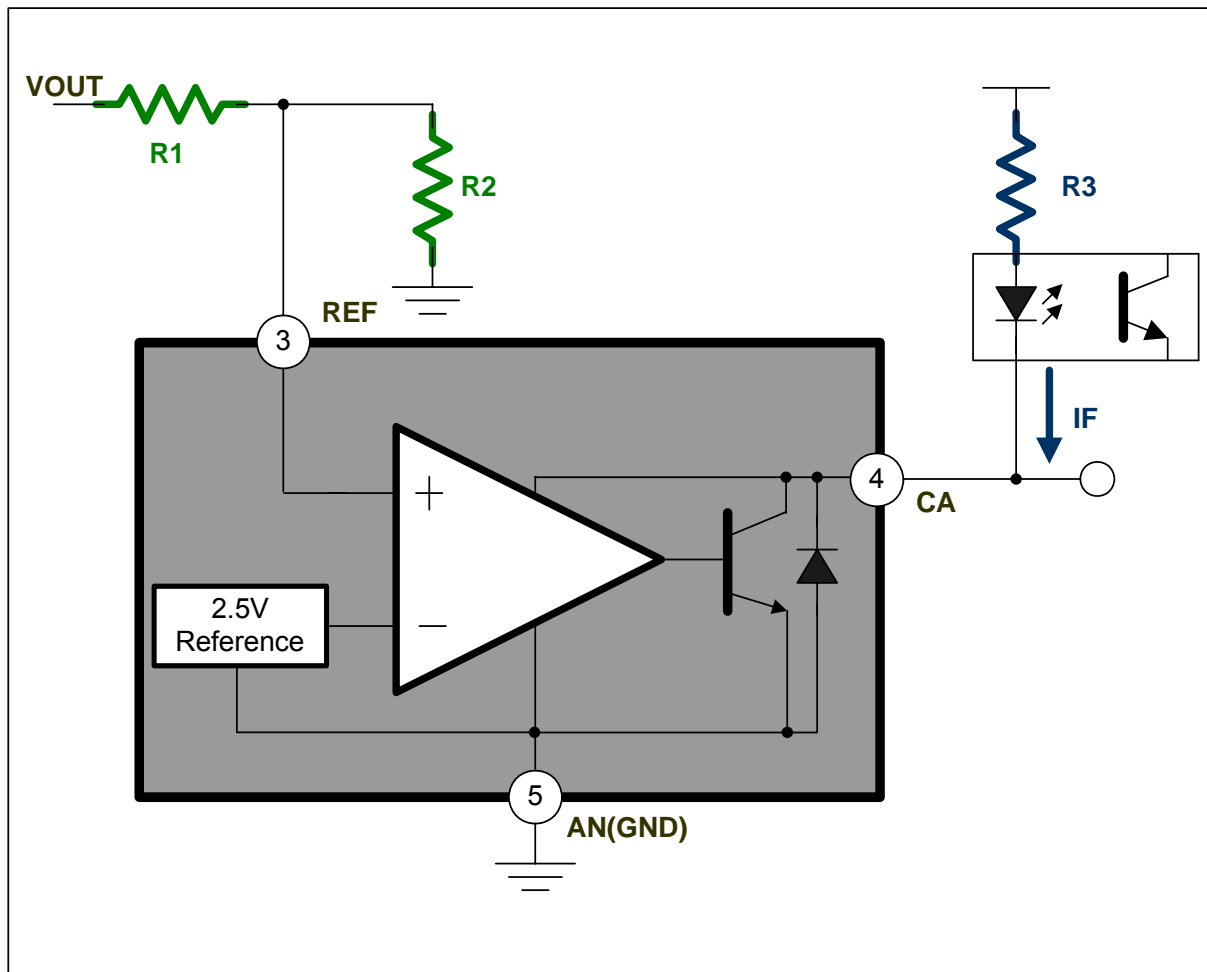
1. The minimum operating voltages of VCC, VIP and VIN are 2.7V, if these supply voltages are low than 2.7V, the transfer function at current shunt output ( $CSO$ ) of **FP133** isn't correct.
2. Don't force a VIN voltage that is over 15V than VIP, this condition would generate a leakage current and an incorrect voltage at **FP133** output.

## Voltage Shunt Regulator

The figure shows the **FP133** voltage shunt regulator, it has an internal reference **2.5V** connecting to the IN- of comparator and high cathode current sink ability for photo-coupler driving.

The VOUT formula is shown below:

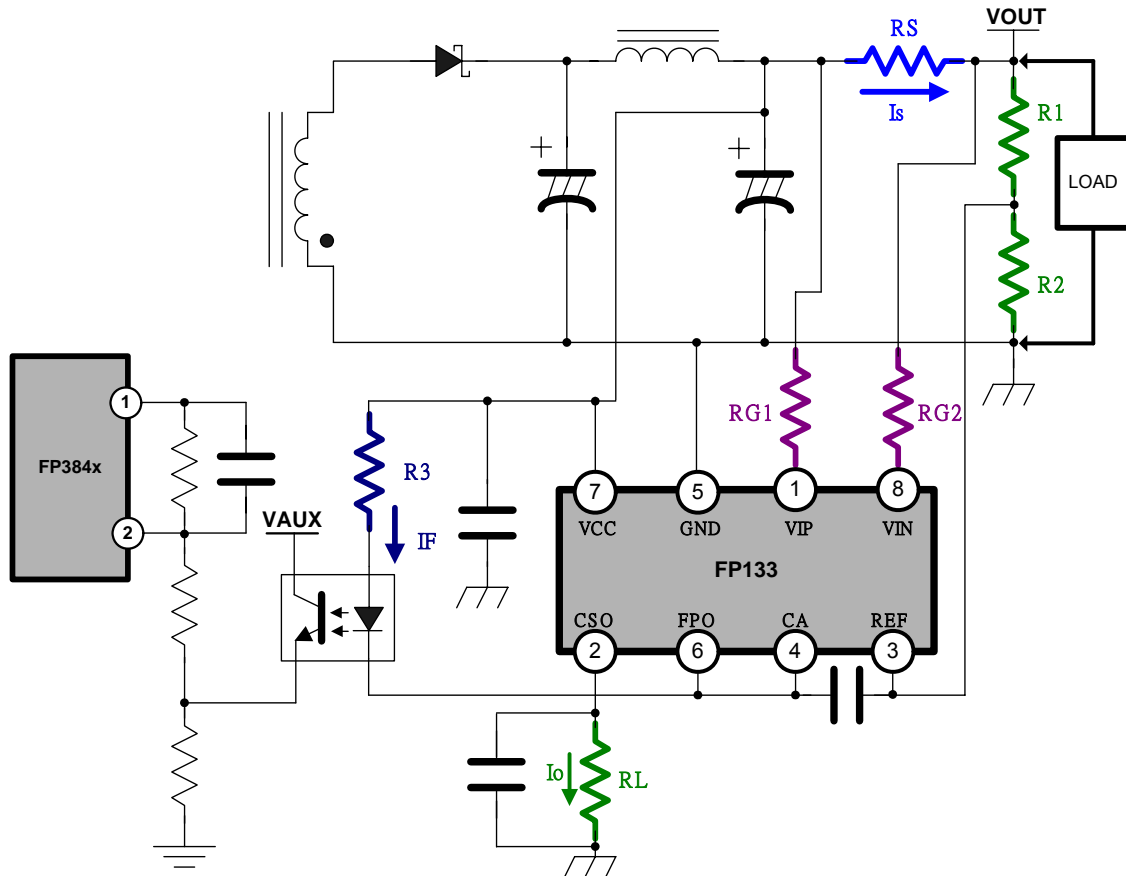
$$V_{OUT} = \left(1 + \frac{R1}{R2}\right) \times 2.5V \quad \text{--- (4)}$$



## NOTE

1. To connecting a compensation network between CA and REF pins avoids output voltage has a high ring during light loading or transient.
2. AN pin must tie to IC GND pin for the same ground level.
3. R3 is selectable for dynamic loading feedback.

## APPLICATION NOTE



### Adaptor secondary Voltage Feedback / Over Loading Protection circuit

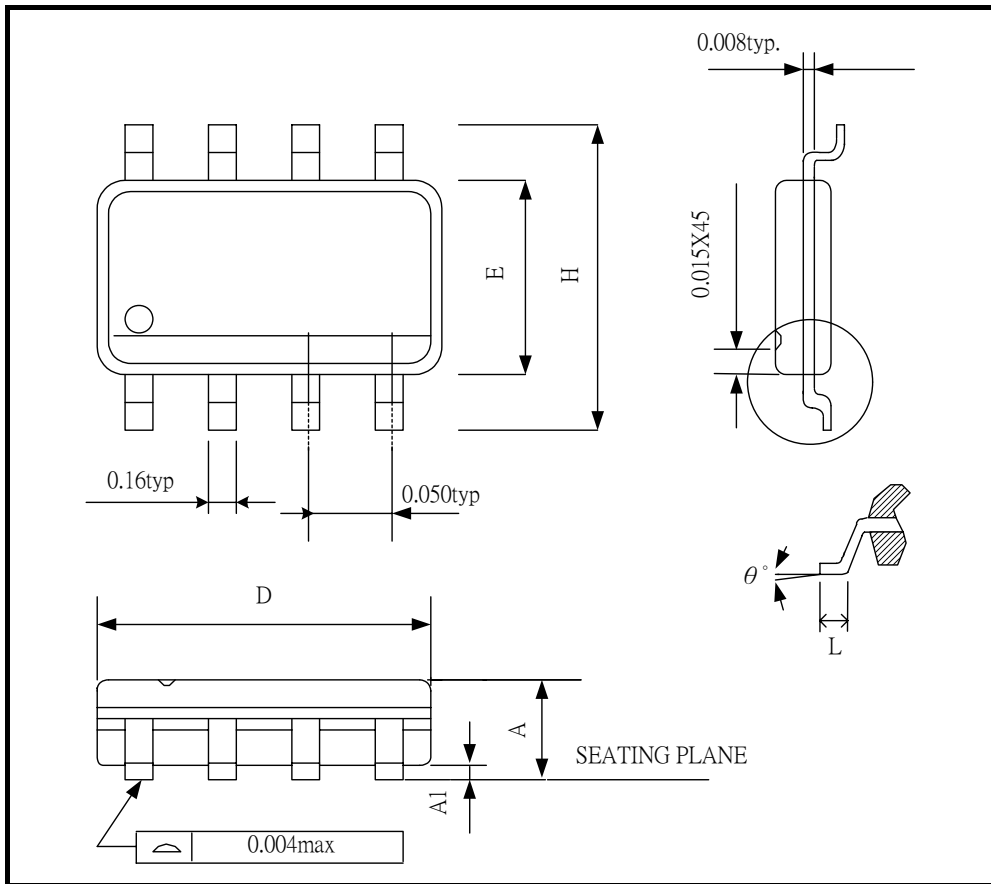
The above circuits are a simple application for AC/DC adaptor over loading protection (OLP) function with output voltage feedback.

For example, when load current ( $I_S$ ) increases, the **FP133** CSO voltage would increase as formula (2) until the internal shutdown comparator IN- connecting to CSO pin is higher than **1.25V** reference, that has a sink current ( $I_{OUT}$ ) flowing through the photo-coupler, and **FP384x** PWM IC will change the NMOS drive terminal to a minimum duty cycle current limitation for secondary over current protection, and primary side auxiliary voltage can not maintain the **FP384x** power supply, the **FP384x** will be shutdown until AC line start-up voltage is re-start the PWM IC.

The **FP133** has a voltage shunt regulator would response the output voltage change with R1 and R2, the feedback current will flow from photo-coupler to **FP133** CA pin (pin4), and generate the feedback signal voltage to PWM IC **FP384x**.

## PACKAGE OUTLINE

SOP8



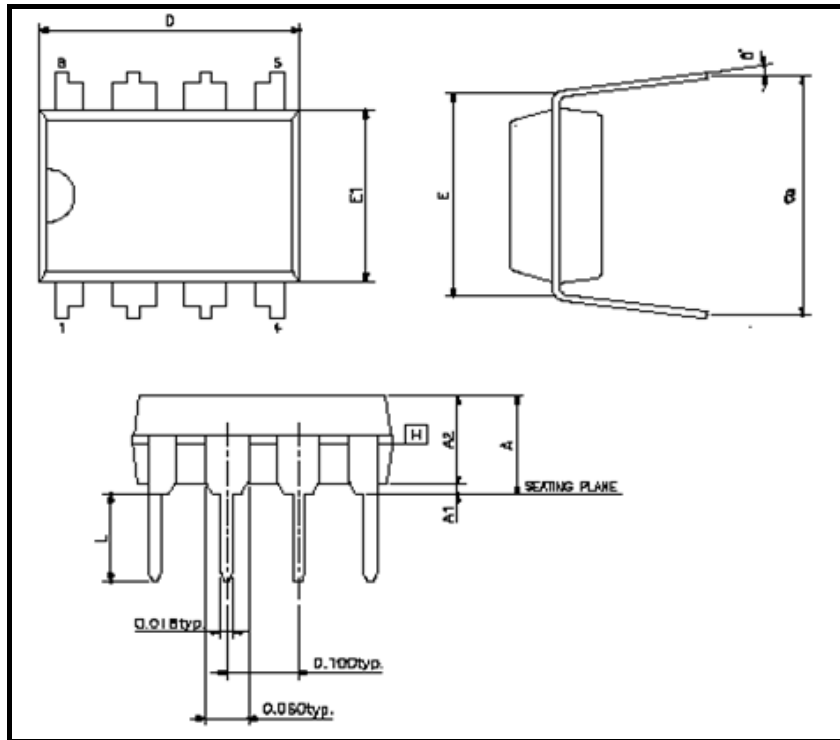
SYMBOLS	MIN	MAX
A	0.053	0.069
A1	0.004	0.010
D	0.189	0.196
E	0.150	0.157
H	0.228	0.244
L	0.016	0.050
$\theta^\circ$	0	8

UNIT:INCH

### NOTE:

1. JEDEC OUTLINE:MS-012 AA ◦
2. DIMENSIONS "D" DOES NOT INCLUDE MOLD FLASH,PROTRUSIONS OR GATE BURRS.MOLD FLASH,PROTRUSIONS AND GATE BURRS SHALL NOT EXCEED .15mm (.06in) PER SIDE ◦
3. DIMENSIONS "E" DOES NOT INCLUDE INTER-LEAD FLASH,OR PROTRUSIONS INTER-LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED .25mm (.0.10in) PER SIDE ◦

PDIP8



SYMBOLS	MIN.	NOR.	MAX.
A	-	-	0.210
A1	0.015	-	-
A2	0.125	0.130	0.135
D	0.355	0.365	0.400
E	0.300 BSC.		
E1	0.245	0.250	0.255
L	0.115	0.130	0.150
e <sub>B</sub>	0.335	0.355	0.375
θ	0	7	15

UNIT:INCH

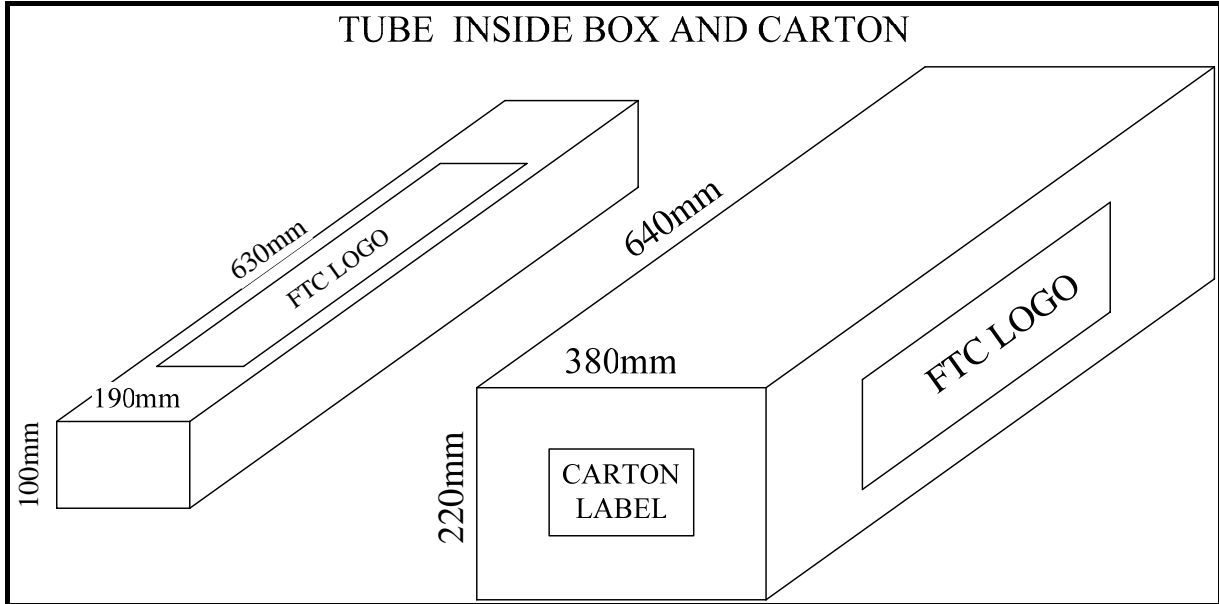
**Note:**

1. JEDEC OUTLINE:MS-001 BA ◦
2. "D"、"E1" DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCH ◦
3. e<sub>B</sub> IS MEASURED AT THE LEAD TIPS WITH THE LEADS UNCONSTRAINED ◦
4. POINTED OR ROUNDED LEAD TIPS ARE PREFERRED TO EASE INSERTION ◦
5. DISTANCE BETWEEN LEADS INCLUDING DAM BAR PROTRUSIONS TO BE .005 INCH MINIMUM ◦
6. DATUM PLANE [H] COINCIDENT WITH THE BOTTOM OF LEAD, WHERE LEAD EXITS BODY ◦

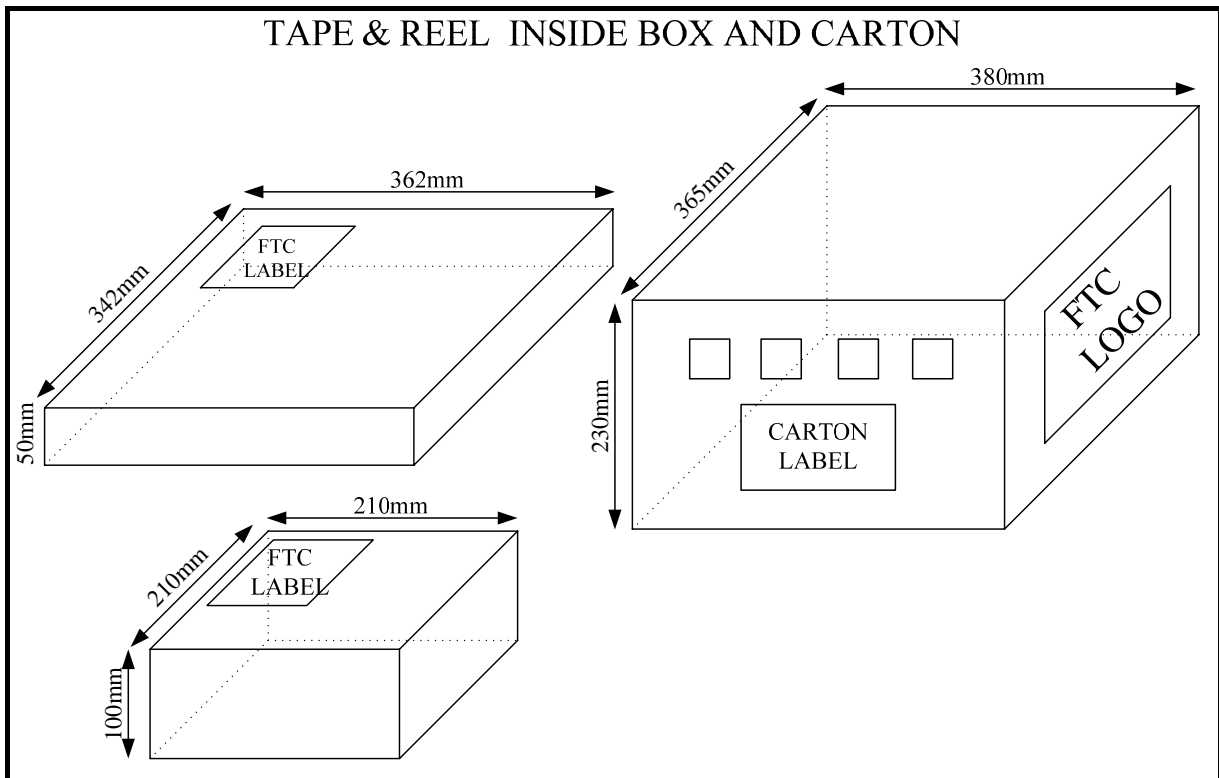
**PACKING SPECIFICATIONS**

**BOX & CARTON DIMENSION**

**PDIP8 & SOP8**



**SOP8 & SOT-25**



**PACKING QUANTITY SPECIFICATIONS**

SOP8	PDIP8
2500 EA / REEL	60 EA/TUBE
1 REELS / INSIDE BOX	60 TUBES / INSIDE BOX
4 INSIDE BOXES / CARTON	4 INSIDE BOXES / CARTON

**LABEL SPECIFICATIONS****TAPPING & REEL**

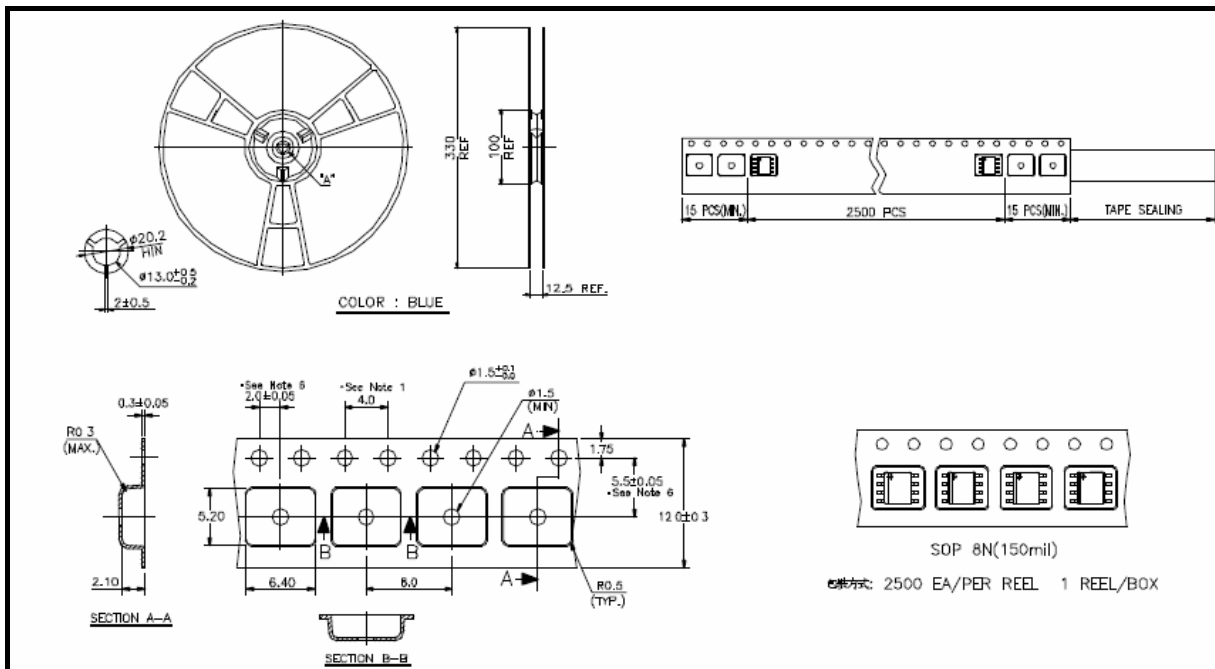
Feeling Technology Corp	
Product:FP133DR-LF	
Lot NO: A3311CXX-L	
D/C: 6Xx-XXL	無鉛 Lead Free
Q`ty: 2500	

**CARTON**

Feeling Technology Corp	
Product Type: FP133DR-LF	
Lot No: A3311CXX-L	
Date Code: 6Xx-XXL	
Package Type:SOP8	
Marking Type:Laser	
Total Q`ty: 10,000	無鉛 Lead Free

## CARRIER TAPE AND REEL DIMENSIONS

### SOP8



#### Note:

1. 10 SPROCKET HOLE PITCH CUMULATIVE TOLERANCE 0.2mm ◦
2. COMBER NOT TO EXCEED 1mm IN 100mm ◦
3. MATERIAL:ANTI-STATIC BLOCK ADVANTEK POLYSTYRENE ◦
4. A<sub>0</sub> AND B<sub>0</sub> MEASURED ON A PLANE 0.3mm ABOVE THE BOTTOM OF THE POCKET ◦
5. K<sub>0</sub> MEASURED FROM A PLANE AN THE INSIDE BOTTOM OF THE POCKET TO THE TOP SURFACE OF THE CARRIER ◦
6. POCKET POSITION RELATIVE TO SPROCKET HOLE MEASURED AS TRUE POSITION OF POCKET , NOT POCK